# Recent Developments in Gain Scheduling Control

by

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#### Outline

- Introduction, definition and motivation.
- Classifications of gain-scheduling control techniques.
- Mathematical descriptions of linear time varying systems.
- Parameter dependent systems and control.
- Systems with linear fractional dependence in the parameters (LFT-systems).
- Linear parameter varying systems with induced quadratic performance.
- Summary.

## What is gain scheduling control?

- Originally a control scheme to counteract nonlinear variations in the steady-state process gain.
- A gain scheduling controller is a parameterized set of linear controllers. In tion is scheduled (determined) according to the parameter operation the parameter is measured (available) and the controller in ac-
- Example: Time varying PID control, where  $K_p$  (may also include  $T_i$  and  $T_d$ ) is tabulated as a function of operating conditions

4

 $ho_1$ 

# Steps in the design of a gain scheduling controller.

- 1) Select a set of stationary operating points.
- 2) Design the controllers for each operating point.
- 3) Design the scheduling algorithm.
- 4) Implement the parameter-scheduled controller.

### Motivating examples

- Mach number (static air pressure and velocity).

1) Gain scheduling control of airplane. Scheduling variables:

- Altitude (height above sea level).
- 2) Application of gain scheduling control in chemical process control.



### Adaptive controller? y linearly

# Classification of gain-scheduling control techniques

### Adam Lagerberg (1996):

- Linear Parameter Varying (LPV) approach.
- Gain-scheduling with Linear Fractional Transformations (LFT).
- Extended linearization and linearization families.
- The D-method.

In this presentation only the LFT and the LPV approaches will be considered.

# Three types of "linear" system descriptions

Linear Time Invariant (LTI) systems

$$\begin{bmatrix} \dot{x} \\ e \end{bmatrix} = \begin{bmatrix} A & B \\ C & D \end{bmatrix} \begin{bmatrix} x \\ d \end{bmatrix}$$

Linear Parameter Varying (LPV) systems

$$\begin{bmatrix} \dot{x} \\ e \end{bmatrix} = \begin{bmatrix} A(\rho) & B(\rho) \\ C(\rho) & D(\rho) \end{bmatrix} \begin{bmatrix} x \\ d \end{bmatrix}$$

Linear Time Varying (LTV) systems

$$\begin{bmatrix} \dot{x} \\ e \end{bmatrix} = \begin{bmatrix} A(t) & B(t) \\ C(t) & D(t) \end{bmatrix} \begin{bmatrix} x \\ d \end{bmatrix}$$

Note, a LPV-system becomes a:

- 1) LTI-system for  $\rho = \text{const.}$  and
- 2) LTV-system for  $\rho = \rho(t)$  (along a time-varying trajectory).

## Parameter-Dependent Systems

## Parameter-Dependent linear plants:

$$\begin{bmatrix} \dot{x} \\ e \\ y \end{bmatrix} = \begin{bmatrix} A(\rho) & B_1(\rho) & B_2(\rho) \\ C_1(\rho) & D_{11}(\rho) & D_{12}(\rho) \\ C_2(\rho) & D_{21}(\rho) & D_{22}(\rho) \end{bmatrix} \begin{bmatrix} x \\ d \\ u \end{bmatrix}$$

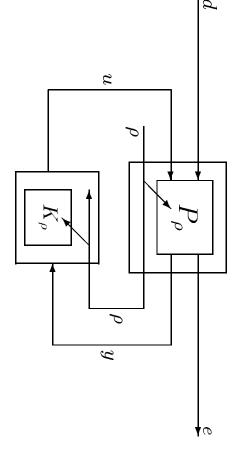
### The parameter $\rho$ :

- is time-varying, i.e.  $\rho(t)$ ,
- takes values in a compact set  $\mathcal{P}$ , and
- there are known bounds on  $\dot{
  ho}$  (which may, or may not be exploited)

#### For control:

- The time variations are not known in advance.
- The parameter values  $\rho(t)$  are measured in real-time with sensors.

# Parameter-Dependent Control of Parameter-Dependent Systems



- Parameter dependent controller  $K_{
  ho}$  which processes
- b and
- $\rho$  nonlinearly

Optimizing closed-loop performance with respect to parameter variations.

## Parameter-Dependent Systems

# Consider two types of Parameter-Dependent systems

- Linear Parameter Varying (LPV), and
- Linear Fractional Transformation (LFT) models.

#### **Distinctions**

- The allowable dependence that the state-space data has on the parameters.
- the analysis To which extent information about the parameter's variation are exploited in
- The different techniques used to analyze the systems.

### Obvious questions

### Natural to look into:

- Stability.
- Stabilizability and Detectability.
- Parameterization of all stabilizing controllers.
- Choosing K to optimize performance.

# Parameter-Dependent Systems: Stability Definitions

#### Two facts:

- Stability of LTV-system is well characterized.
- When evaluating along a allowable trajectory, the P-D system becomes an LTV-system.

LFT and LPV systems, two different approaches to stability tests:

- LFT-systems: Structured Small-Gain theorems.
- LPV-systems: Parameter dependent Lyapunov functions,

# P-D Systems: Stabilizability and Detectability

Stabilizability of  $(A(\rho), B(\rho))$ : If there exists a P-D state-feedback u(t) = $F(\rho)x(t)$ , such that:

$$\dot{x} = A(\rho)x(t) + B(\rho)F(\rho)x(t)$$

is stable for all possible parameters  $ho \in \mathcal{P}$ 

Detectability of  $(C(\rho), A(\rho))$ : If there exists a P-D output to state-feedback  $L(\rho)$ , such that:

$$\dot{x} = A(\rho)x(t) + L(\rho)C(\rho)x(t)$$

is stable for all possible parameters  $ho\in\mathcal{P}$ 

## P-D Systems: All "stabilizing" controllers

Extension to Youla parameterization (Wei Min Lu) for

- the two types of P-D systems
- with their corresponding notations on P-D stability.

controlled by their corresponding P-D controllers

Output feedback stabilization:

Given an open-loop P-D system

$$\begin{bmatrix} \dot{x} \\ y \end{bmatrix} = \begin{bmatrix} A(\rho) & B(\rho) \\ C(\rho) & D(\rho) \end{bmatrix} \begin{bmatrix} x \\ u \end{bmatrix}$$

- if and only if There exists a finite-dimensional P-D controller that stabilizes the system
- 1) The pair  $(A(\rho),B(\rho))$  is stabilizable
- 2) The pair  $(C(\rho), A(\rho))$  is detectable
- Usual observer structure.

### LFT-systems

#### Given:

Real-parameters  $\delta_i$  which may be repeated, give rise to:

$$\left\{ \mathcal{D}_{\mathcal{S}} \triangleq \operatorname{diag}\left\{\delta_{1}I_{s_{1}}, \dots, \delta_{f}I_{s_{f}}\right\} : \delta_{i} \in \mathbb{R} \right\} \subset \mathbb{R}^{s \times s}, \quad \mathcal{S} \triangleq \left(s_{1}, s_{2}, \dots, s_{f}\right)$$

System matrix M, partitioned according to:

$$\begin{bmatrix} \dot{x}(t) \\ e(t) \\ \alpha(t) \end{bmatrix} = \begin{bmatrix} M_{11} & M_{12} & M_{22} \\ M_{21} & M_{22} & M_{23} \\ M_{31} & M_{32} & M_{33} \end{bmatrix} \begin{bmatrix} x(t) \\ d(t) \\ \beta(t) \end{bmatrix}, \quad \beta(t) = \Delta(t)\alpha(t)$$

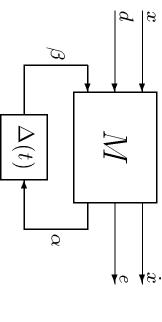
The LFT-system  $G_{\Delta}$  is described by  $G_{\Delta}=\mathcal{F}_l(M,\Delta(t))$ , with state-space equations

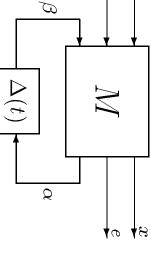
$$\begin{bmatrix} \dot{x}(t) \\ e(t) \end{bmatrix} = \underbrace{\left\{ \begin{bmatrix} M_{11} & M_{12} \\ M_{21} & M_{22} \end{bmatrix} + \begin{bmatrix} M_{13} \\ M_{23} \end{bmatrix} \Delta(t)(I - M_{33}\Delta(t))^{-1} \begin{bmatrix} M_{31} & M_{32} \end{bmatrix} \right\} \begin{bmatrix} x(t) \\ d(t) \end{bmatrix}}_{G_{\Delta}}$$

where allowable piecewise continues  $\Delta(t)$  trajectories satisfy:

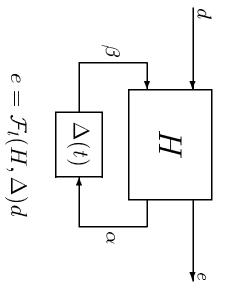
$$\Delta(t) = \mathcal{D}_{\mathcal{S}}(\delta(t)), \quad |\delta_i(t)| \leq 1, \quad \text{however, no restrictions on } \delta.$$

### LFT-systems graphically





$$\begin{bmatrix} \dot{x}(t) \\ e(t) \end{bmatrix} = \mathcal{F}_l(M, \Delta(t)) \begin{bmatrix} x(t) \\ u(t) \end{bmatrix}$$



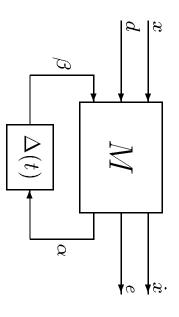
where H(s) is the 2-input, 2-output transfer function such that

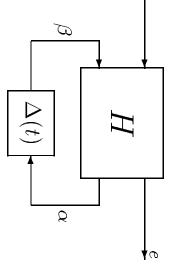
$$\begin{bmatrix} e \\ \alpha \end{bmatrix} = \begin{bmatrix} H_{11}(s) & H_{12}(s) \\ H_{21}(s) & H_{22}(s) \end{bmatrix} \begin{bmatrix} d \\ \beta \end{bmatrix}$$

H(s) given in terms of M is

$$H(s) = \begin{bmatrix} M_{22} & M_{23} \\ M_{32} & M_{33} \end{bmatrix} + \begin{bmatrix} M_{21} \\ M_{31} \end{bmatrix} (sI - M_{11})^{-1} \begin{bmatrix} M_{12} & M_{13} \end{bmatrix}$$

# Robust stability condition for LFT-systems with linear time varying \( \Delta \)





The LFT-system (M, S) is stable if

- 1)  $M_{11}$  is Hurwitz, and
- 2) there exists a constant matrix  $Z = Z^T > 0$  of the form

$$Z=\mathrm{diag}\{Z_1,Z_2,\ldots,Z_f\}, \quad \text{with} \quad Z_i\in\mathbb{R}^{s_i imes s_i}, \quad \text{such that} \quad Z\Delta(t)=\Delta(t)Z_t\}$$

satisfying

$$\left\| Z^{\frac{1}{2}} H_{22} Z^{-\frac{1}{2}} \right\|_{\infty} < 1$$

If so, small-gain argument verify exponential stability of

$$\dot{x}(t) = \{M_{11} + M_{13}\Delta(t)(I - M_{33}\Delta(t))^{-1}M_{31}\}x(t)$$

# Scaled bounded real lemma (Gahinet and Apkarian, 1994)

With  $H_{22} = C(sI - A)^{-1}B + D$ , then the following two statements are equivalent:

1) A is stable and there exists a constant diagonal matrix  $Z=Z^T>0$ 

$$Z=\mathrm{diag}\{Z_1,Z_2,\ldots,Z_f\},$$
 with  $Z_i\in\mathbb{R}^{s_i},$  such that  $Z\Delta(t)=\Delta(t)Z$ 

satisfying

$$\left\| Z^{\frac{1}{2}} H_{22} Z^{-\frac{1}{2}} \right\|_{\infty} < 1$$

2) There exist solutions  $X = X^T > 0$  and  $Z = Z^T > 0$  (with the structure give above) such that  $\int A^T X + XA \quad XB$ < 0

$$\begin{bmatrix} A & A & A & A & A & A \\ B^T X & -Z & I \\ C & D & -A \end{bmatrix}$$

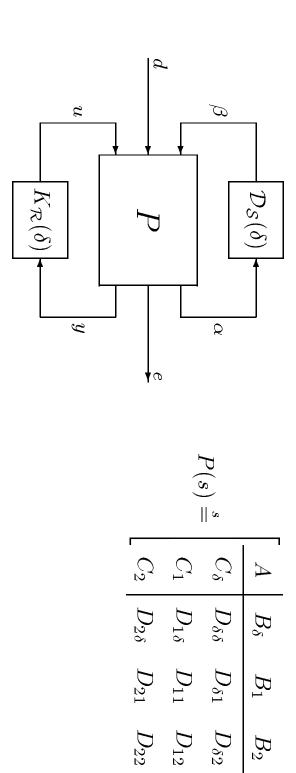
is fulfilled.

### LFT closed-loop system

Define time invariant rational transfer function matrix P(s)

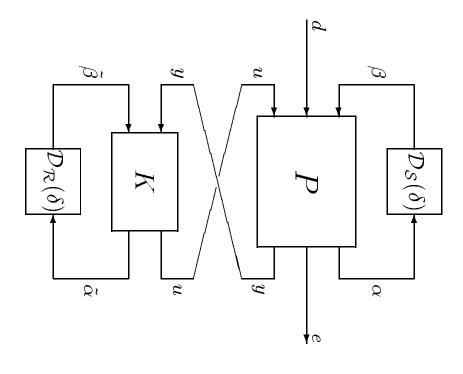
$$\begin{bmatrix} \alpha \\ e \\ y \end{bmatrix} = \begin{bmatrix} P_{11} & P_{12} & P_{13} \\ P_{21} & P_{22} & P_{23} \\ P_{31} & P_{32} & P_{33} \end{bmatrix} \begin{bmatrix} \beta \\ d \\ u \end{bmatrix}$$

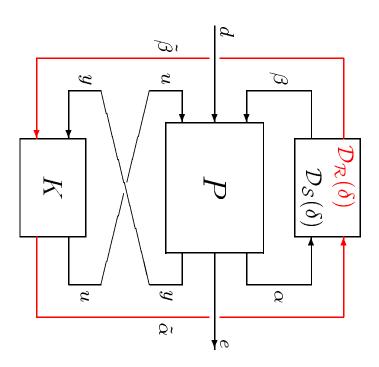
Interconnection of u and y, i.e.  $u = K_{\mathcal{S}}(\delta)y$ 



The time varying parameters  $\delta$  enter both in  $\mathcal{D}_{\mathcal{S}}(\delta)$  and in  $K_{\mathcal{R}}(\delta)$ !

### LFT closed-loop system





### LFT closed-loop system

Closed-loop transfer function is:

$$T(P, K, \mathcal{D}_{\mathcal{S}}(\delta), \mathcal{D}_{\mathcal{R}}(\delta)) = \mathcal{F}_{l}(\mathcal{F}_{u}(P, \mathcal{D}_{\mathcal{S}}(\delta)), \mathcal{F}_{l}(K, \mathcal{D}_{\mathcal{R}}(\delta)))$$

Define  $P_a(s)$  $ilde{eta}_{ ilde{eta}}^{ ilde{oldsymbol{y}}}$ Ω̃  $= \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & P_{11} & P_{12} & P_{13} \\ 0 & P_{21} & P_{22} & P_{23} \\ 0 & P_{31} & P_{32} & P_{33} \\ I_r & 0 & 0 & 0 \end{bmatrix}$ 

Closed-loop transfer function in terms of  $P_a(s)$ :

$$T(P_a, K, \mathcal{D}_{\mathcal{S}}(\delta), \mathcal{D}_{\mathcal{R}}(\delta)) = \mathcal{F}_u\left(\mathcal{F}_l(P_a(s), K(s)), \begin{bmatrix} \mathcal{D}_{\mathcal{R}}(\delta) \\ \mathcal{D}_{\mathcal{S}}(\delta) \end{bmatrix}\right)$$

block uncertainty  $\left[ egin{array}{c} {\cal D}_{\cal R}(\delta) \end{array} 
ight]$ performance control problem with the nominal plant  $P_a$ , and with repeated rea The LFT gain-scheduling control problem can be treated as a classical robust  $\mathcal{D}_{\mathcal{S}}(\delta)$  ] .

## Induced $L_2$ performance in LFT-systems

Given a time varying vector e(t), then the  $L_2$ -norm of e(t) is defined by

$$\|e(t)\|_{2} \triangleq \left\{ \int_{0}^{\infty} \|e(\tau)\|^{2} d\tau \right\}^{\frac{1}{2}} < \infty$$

Induced L<sub>2</sub> performance metric

$$\max_{\mathsf{allowable}\delta}\max_{d}\frac{\|G_{\Delta}d\|_2}{\|d\|_2}$$

i.e. take the "worst-case" (over all parameter trajectories) induced  $L_2$  gain

Induced  $L_2$  performance bound  $\gamma$  ( $L_2$ -gain)

$$\int_0^T e^T(\tau)e(\tau)d\tau \le \gamma^2 \int_0^T d^T(\tau)d(\tau)d\tau \quad \forall T \ge 0$$

### LFT synthesis problem

#### Given:

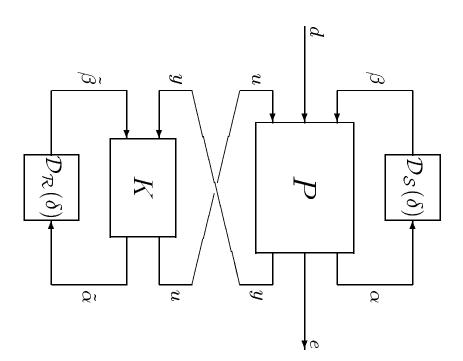
1) Finite dimensional linear P

$P(s) \stackrel{s}{=}$		
$C_{s}$ $C_{1}$	A	
$D_{\delta\delta} \ D_{1\delta} \ D_{2\delta}$	$B_\delta$	
$D_{\delta 1} \ D_{11} \ D_{21}$	$B_1$	
$egin{array}{c} D_{\delta 2} \ D_{12} \ D_{22} \ \end{array}$	$B_2$	

2) Vector of integers S describing the plant's parameter structure.

### Find (if possible):

- 1) Finite dimensional linear K.
- 2) Vector of integers  $\mathcal{R}$ , describing the controller's parameter structure.



### LFT synthesis setup

- $\Delta(t) \in \mathcal{D}_{\mathcal{S}}(\delta)$ .
- Let  $Z_{\Delta}$  denote the set of matrices such that for  $Z \in Z_{\Delta}$  so that  $Z\Delta = \Delta Z$
- Let  $\mathcal{N}_X$  and  $\mathcal{N}_Y$  be bases for the null spaces of

$$egin{bmatrix} [B_2^T & D_{\delta 2}^T & D_{12}^T & 0 \end{bmatrix}$$
 and  $egin{bmatrix} C_2 & D_{2\delta} & D_{21} & 0 \end{bmatrix}$ 

ullet Let  $X_i \in \mathbb{R}^{s_i imes s_i}$ ,  $Y_i \in \mathbb{R}^{s_i imes s_i}$  and define

$$\widehat{X} = \operatorname{diag}\{X_1, X_2, \dots, X_f\}$$
 and  $\widehat{Y} = \operatorname{diag}\{Y_1, Y_2, \dots, Y_f\}$ 

then  $\widehat{X},\widehat{Y}\in Z_{\Delta}$  .

Define

$$\widehat{B}_1 = \begin{bmatrix} B_\delta & B_1 \end{bmatrix}, \quad \widehat{C}_1 = \begin{bmatrix} C_\delta \\ C_1 \end{bmatrix} \quad \text{and} \quad \widehat{D}_{11} = \begin{bmatrix} D_{\delta\delta} & D_{\delta1} \\ D_{1\delta} & D_{11} \end{bmatrix}$$

### LFT synthesis solution

bound  $\gamma$ , if there exist matrices: The LFT gain-scheduling control problem is solvable with closed-loop performance

$$X_0 = X_0^T \in \mathbb{R}^{n \times n}, \quad Y_0 = Y_0^T \in \mathbb{R}^{n \times n}, \quad \widehat{X} = \widehat{X}^T \in Z_\Delta \quad \text{and} \quad \widehat{Y} = \widehat{Y}^T \in Z_\Delta$$

such that

$$\mathcal{N}_{X} \begin{bmatrix} AX_{0} + X_{0}A^{T} & X_{0}\widehat{C}_{1}^{T} & \widehat{B}_{1} \\ \widehat{C}_{1}X_{0} & -\gamma \begin{bmatrix} \widehat{Y} & 0 \\ 0 & I \end{bmatrix} & \widehat{D}_{11} \\ \widehat{B}_{1}^{T} & \widehat{D}_{11}^{T} & -\gamma \begin{bmatrix} \widehat{X} & 0 \\ 0 & I \end{bmatrix} \end{bmatrix} \mathcal{N}_{Y} < 0 \tag{1}$$

$$\mathcal{N}_{Y} \begin{bmatrix} A^{T}Y_{0} + Y_{0}A & Y_{0}\widehat{B}_{1} & -\gamma \begin{bmatrix} \widehat{X} & 0 \\ 0 & I \end{bmatrix} \\ \widehat{B}_{1}^{T}Y_{0} & -\gamma \begin{bmatrix} \widehat{X} & 0 \\ 0 & I \end{bmatrix} & \widehat{D}_{11}^{T} \\ \widehat{C}_{1} & \widehat{D}_{11} & -\gamma \begin{bmatrix} \widehat{Y} & 0 \\ 0 & I \end{bmatrix} \end{bmatrix} \mathcal{N}_{X} < 0 \tag{2}$$

$$\begin{bmatrix} X_{0} & I \\ I & Y_{0} \end{bmatrix} \ge 0 \text{ and } \begin{bmatrix} \widehat{X} & I \\ I & \widehat{Y} \end{bmatrix} \ge 0 \tag{3}$$

## LFT synthesis solution summary

Sufficient condition for existence of LFT-controller such that the closed loop system has performance level  $< \gamma$ :

if and only if there are matrices  $X=X^{T}>0$  and  $Y=Y^{T}>0$  with structure

$$X = \text{diag}\{X_0, X_1, \dots, X_f\}, \quad Y = \text{diag}\{Y_0, Y_1, \dots, Y_f\},$$

satisfying

$$\mathsf{AMII}_1(X,\gamma) < 0, \quad \mathsf{AMII}_2(Y,\gamma) < 0 \quad \text{and} \quad \begin{bmatrix} X_i & I \\ I & Y_i \end{bmatrix} \geq 0$$

ullet  $\gamma$ -suboptimal controller of order k if

$$\operatorname{rk}(I - X_0 Y_0) \le k$$

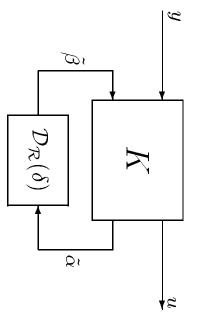
Some comments:

- The linear matrix inequalities are finite dimensional.
- Robust numerical methods for solving the affine linear matrix inequalities exist.

### LFT-controller

## Parameter dependent controller:

- $\mathcal{R} = \mathcal{S}$ , i.e. the controller's parameter structure is the same as the plant's parameter structure
- The LTI part of the LFT controller K, is reconstructed from X and Y, i.e. the solutions to the affine matrix inequalities
- Implementation:



With  $\delta$  given (measured):

$$K_{\mathcal{R}}(\delta) = \mathcal{F}_l(K, \mathcal{D}_{\mathcal{R}}(\delta))$$
  
 $u = K_{\mathcal{R}}(\delta)y$ 

• The control action u is linear in y, nonlinear in  $\delta$ .

## LFT extensions and remarks

- Results on LFT synthesis derived independently by (Packard, 1997):
- Andy Packard, Greg Becker and Fen Wu, and
- Pierre Apkarian and Pascal Gahinet.
- But note, similar ideas have been proposed by Lu and Doyle (1992,1995).
- parameters  $\delta$ The Small-Gain LFT test may be conservative due to the realness of the
- Helmersson has generalized the ideas to exploit the realness of the parameters (upper bound on  $\mu$  for repeated real parameters).
- The synthesis conditions are still AMI's in block diagonal form
- Controller structure is still a LFT of a fixed LTI-system
- by Braatz and Morari (1997). An improved upper bound on  $\mu$  for repeated real parameters is also given
- the rate variations in the parameters  $\delta$ The results may also be conservative since they do not take into account

### LPV-systems

Parameter set,  $\mathcal{P}\subset \mathbb{R}^f$ , LPV-system  $G_
ho$  on state-space form:

$$\begin{bmatrix} \dot{x} \\ e \\ y \end{bmatrix} = \begin{bmatrix} A(\rho) & B_1(\rho) & B_2(\rho) \\ C_1(\rho) & D_{11}(\rho) & D_{12}(\rho) \\ C_2(\rho) & D_{21}(\rho) & 0 \end{bmatrix} \begin{bmatrix} x \\ d \\ u \end{bmatrix}$$

where a allowable trajectory  $\rho(\cdot)$  satisfy:

- $\rho(t) \in \mathcal{P}$ , for all t and
- for each  $\rho_i$  and all t,  $\underline{\nu}_i(\rho(t)) \leq \dot{\rho}_i(t) \leq \bar{\nu}_i(\rho(t))$ , i.e. bounded rate of varia-

# LPV-systems: control problem formulation

The gain-scheduling output feedback controller  $K(\rho)$  is given by:

$$\begin{bmatrix} \dot{x}_K \\ u \end{bmatrix} = \begin{bmatrix} A_K(\rho, \dot{\rho}) & B_K(\rho, \dot{\rho}) \\ C_K(\rho, \dot{\rho}) & D_K(\rho, \dot{\rho}) \end{bmatrix} \begin{bmatrix} x_K(t) \\ y \end{bmatrix}$$

Induced  $L_2$  performance bound  $\gamma$ 

$$\int_0^T e^T(\tau)e(\tau)d\tau \le \gamma^2 \int_0^T d^T(\tau)d(\tau)d\tau \quad \forall T \ge 0$$

• Lyapunov function  $V(x_{cl}, P(\rho)) = x_{cl}P(\rho)x_{cl}$ 

# LPV-systems: Basic characterization (Apkarian and Adams, 1998)

and a performance bound  $\gamma$ , whenever there exist P-D matrices  $X=X^T>0$  and  $Y=Y^T>0$  and a P-D quadruple  $(A_K,B_K,C_K,D_K)$ , such that There exists a gain-scheduling output-feedback controller enforcing internal stability

$$\begin{bmatrix} \dot{X} + XA + \hat{B}_K C_2 + (\star) & \star & \star & \star \\ \hat{A}_K^T + A + B_2 D_K C_2 & -\dot{Y} + AY + B_2 \hat{C}_K + (\star) & \star & \star \\ (XB_1 + \hat{B}_K D_{21})^T & (B_1 + B_2 D_K D_{21})^T & -\gamma I & \star \\ (C_1 + D_{12} D_K C_2 & C_1 Y + D_{12} \hat{C}_K & D_{11} + D_{12} D_K D_{21} & -\gamma I \end{bmatrix} < \begin{bmatrix} X & I \\ I & Y \end{bmatrix} > 0$$

The controller  $(A_K, B_K, C_K, D_K)$  can be reconstructed from:

$$X, \quad Y \quad \mathsf{and} \quad (\hat{A}_K, \hat{B}_K, \hat{C}_K, D_K)$$

# LPV-systems: Projected solvability conditions (Apkarian and Adams, 1998)

and a performance bound  $\gamma$ , whenever there exist P-D matrices  $X=X^T>0$  and  $Y = Y^T > 0$ , such that There exists a gain-scheduling output-feedback controller enforcing internal stability

$$\begin{bmatrix} \mathcal{N}_X & 0 \\ 0 & I \end{bmatrix}^T \begin{bmatrix} \dot{X} + XA + A^TX & XB_1 & C_1^T \\ B_1^TX & -\gamma I & D_{11}^T \\ C_1 & D_{11} & -\gamma I \end{bmatrix} \begin{bmatrix} \mathcal{N}_X & 0 \\ 0 & I \end{bmatrix} < 0$$

$$\frac{V_Y}{0} \begin{bmatrix} 0 \\ I \end{bmatrix}^T \begin{bmatrix} -\dot{Y} + YA^T + AY & YC_1^T & B_1 \\ C_1Y & -\gamma I & D_{11} \\ B_1^T & D_{11}^T & -\gamma I \end{bmatrix} \begin{bmatrix} \mathcal{N}_X & 0 \\ 0 & I \end{bmatrix}$$

$$\begin{bmatrix} X & I \\ I & Y \end{bmatrix} > 0$$

where  $\mathcal{N}_X$  and  $\mathcal{N}_Y$  are bases for the nullspaces of  $\begin{bmatrix} C_2 & D_{21} \end{bmatrix}$  and  $\begin{bmatrix} B_2^T & D_{12}^T \end{bmatrix}$ .

# LPV-systems: Projected solvability conditions (Apkarian and Adams, 1998)

Existence conditions become necessary and sufficient if quadratic stability is imposed through Lyapunov function:

$$V(x_{cl},P(
ho))=x_{cl}^TP(
ho)x_{cl}, \quad ext{with} \quad x_{cl}=\left[egin{array}{c} x \ x_K \end{array}
ight]$$

The controller can be constructed from X and Y along the lines of (Gahinet, 1994).

## LPV-systems: Summary and remarks

- Induced  $L_2$  performance can be tested in terms of two affine linear matrix inequalities (AMI). However, these are dependent on the parameters  $\rho$ .
- This yields a infinite dimensional convex optimization problem
- The suggested solution is to grid the parameter set  $\mathcal{P}$
- Pick a basis for the parameter dependent solutions  $X(\rho)$  and  $Y(\rho)$  to the AMI's.
- Solve the AMI's at the grid points.
- The number of inequalities grow exponentially with the number of parameters.
- The number of inequalities grow linearly with number of grid points.

## LPV-systems: Summary and remarks

If the state-space matrices  $(A(\rho),B(\rho),C(\rho),D(\rho))$  depend in affine manner on the parameters  $\rho$ , i.e.

$$\begin{bmatrix} A(\rho) & B(\rho) \\ C(\rho) & D(\rho) \end{bmatrix} = \begin{bmatrix} A_0 & B_0 \\ C_0 & D_0 \end{bmatrix} + \sum_{i=1}^f \rho_i \begin{bmatrix} A_i & B_i \\ C_i & D_i \end{bmatrix}$$

finite dimensional AMI's become finite dimensional. Then it is sufficient to test the corner points (Apkarian et al., 1995). Then the in-

The controller becomes dependent on  $\dot{
ho}$ . In order to be practically valid, this dependence must be removed, for further details see (Apkarian and Adams, 1998).

# Summary on "recent work" in gain-scheduling control.

- Renewed interest in gain-scheduling control and LPV-systems, due to new which can be applied to LMI's. powerful techniques and computational schemes (interior point methods)
- Some nonlinear control problems can be solved.
- Focuses on analysis and theoretical development rather than ad. hoc. approaches
- Applications using the LFT and LPV techniques start to emerge.
- Number of papers and the focus from academia is increasing.
- Several parallels between gain-scheduling control and model predictive control.

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